# Multimode bolometer development for the Primordial Inflation Explorer (PIXIE) instrument



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# **Outline**

- 1. Introduction and instrument description
- 2. Detector design and fabrication
- 3. Package and readout
- 4. Detector performance
- 5. Conclusions



Introduction and instrument

description

### Introduction

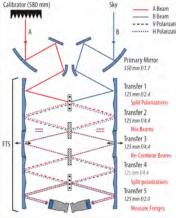
The Primordial Inflation Explorer (PIXIE) [1, 2]

- Space-based polarizing Fourier transform spectrometer (FTS).
- Designed to measure the polarization and intensity spectra of the CMB.
- Multimode "lightbucket" design enables nK-scale sensitivity across 2.5 decades in frequency with just 4 thermistor-based bolometers.
- Like other FTSs [3, 4, 5, 6], PIXIE's design and experimental approach<sup>a</sup> represent a significant departure from imagers often used for CMB measurements. This is especially true for the detectors.
  - Large etendue ( $A\Omega = 4 \text{ cm}^2 \text{ sr}$ ).
  - Handle large optical load (120 pW).
  - Large and mechanically robust absorber structure (30x larger than the spider web bolometers on Planck [7]).
  - Limited sensitivity to particle hits.
  - Sensitive to all optical frequencies of interest (15 GHz 5 THz).
  - Photon-noise limited (NEP  $\leq 1 \times 10^{-16} \ \mathrm{W/\sqrt{Hz}}$ ).



<sup>&</sup>lt;sup>a</sup>See Al Kogut's poster on systematic error mitigation and Dale Fixsen's talk on beams.

# Instrument description



Each focal plane has two polarization-sensitive bolometers mounted back-to-back with their polarization axes orthogonal.

Incident radiation:

$$\vec{E}_{inc} = A\hat{x} + B\hat{y} \tag{1}$$

Measured power:

$$\begin{split} \mathbf{P}_{x}^{L} &= \frac{1}{2} \int \left( A^{2} + B^{2} \right) + \left( A^{2} - B^{2} \right) \cos \left( \frac{4\nu z}{c} \right) d\nu. \\ \mathbf{P}_{y}^{L} &= \frac{1}{2} \int \left( A^{2} + B^{2} \right) + \left( B^{2} - A^{2} \right) \cos \left( \frac{4\nu z}{c} \right) d\nu. \\ \mathbf{P}_{x}^{R} &= \frac{1}{2} \int \left( A^{2} + B^{2} \right) + \left( A^{2} - B^{2} \right) \cos \left( \frac{4\nu z}{c} \right) d\nu. \\ \mathbf{P}_{y}^{R} &= \frac{1}{2} \int \left( A^{2} + B^{2} \right) + \left( B^{2} - A^{2} \right) \cos \left( \frac{4\nu z}{c} \right) d\nu. \end{split}$$

Inverse Fourier transform:

$$\mathbf{S}_{X}^{L}(\nu) = A_{\nu}^{2} - B_{\nu}^{2}.$$

$$\mathbf{S}_{y}^{L}(\nu) = B_{\nu}^{2} - A_{\nu}^{2}.$$

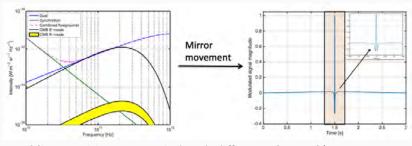
$$\mathbf{S}_{X}^{R}(\nu) = A_{\nu}^{2} - B_{\nu}^{2}.$$

$$\mathbf{S}_{y}^{R}(\nu) = B_{\nu}^{2} - A_{\nu}^{2}.$$
(3)

Signal = small modulated component in a bright ( $\sim 120 \, \mathrm{pW}$ ) background.



# Instrument description



- Mirror position  $z \to \text{optical path difference } \ell \colon z \simeq \ell/4.$
- Mirror velocity v: v = z/(3 sec) = 1.73 mm/sec.
- Optical path difference  $\ell \to \text{interfering radio frequency } \nu \colon \ell = c/\nu .$
- Radio frequency  $\nu \to {\sf Audio}$  (FTS) frequency  $\omega$ :  $\omega = 4\nu v/c$ .
- CMB:  $\omega \lesssim$  15 Hz.
- Dust:  $\omega \lesssim 100$  Hz.

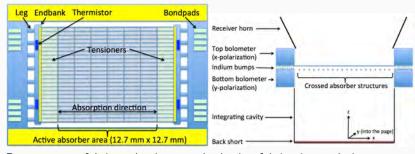
These constraints drive the bolometer bias and bandwidth requirements.

Detectors must be photon noise limited across all FTS frequencies (0  $-100~\rm{Hz})$  under large, near-constant ( $\sim120~\rm{pW})$  optical bias.



Detector design and fabrication

# Detector design - overview



Detectors are fabricated using standard microfabrication techniques.

They consist of three main components:

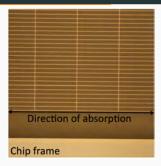
- Absorber structure absorb single linear polarization
- Endbanks measure incident optical power with silicon thermistors
- Frame thermal sink and interface to readout

Each beam's focal plane will consist of two indium bump-hybridized detectors mounted < 20  $\mu\rm m$  apart with their absorbers orthogonal.

 $\rightarrow$  measure orthogonal polarizations of nearly the same electric field.



# Absorber structure - overview



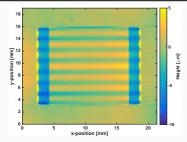


- Consists of a grid of suspended, micromachined, ion implanted silicon wires.
- Wires are degenerately doped to be metallic at all temperatures.
- Effective sheet resistance of the whole structure is 377  $\Omega/\Box$ .
- Absorber area sets low frequency cutoff of instrument (15 GHz); grid spacing (30  $\mu$ m) sets high frequency cutoff (5 THz).
- Wire widths and thicknesses are highly uniform across the array.
  - $\bullet$  Thickness set by starting SOI device layer thickness (1.35  $\mu$ m).
  - Wires are etched to width with an ICP RIE process.



# Absorber structure - mechanical characterization





- Doping induces compressive stress in absorber wires; previous devices had their wires buckle and protrude up to 20  $\mu m$  from the frame.
  - $\rightarrow$  problematic for a hybridized pair of bolometers.
- Detectors subject to vibrations and acoustic excitations at launch.
  - ightarrow need resonant frequencies of absorber structure to be much greater than excitation frequencies of launch.
- $\bullet$  Solution: deposit highly tensile  ${\rm Al_2O_3}$  film on absorbers outside of active optical region.
  - ightarrow Fabricated absorbers are flat and expected to oscillate with amplitudes of < 0.4  $\mu m$  rms during launch.

## **Endbanks - overview**



- Consists of a gold bar for thermalization and two doped silicon thermistors on a crystalline silicon membrane.
- The gold bar also sets the heat capacity of the endbank.
- Endbank is formed from the device layer of the SOI substrate.
- Endbanks are connected to the chip frame through eight silicon legs.
- Thermistors are doped to operate below metal-insulator transition. Electron transport mechanism is variable range hopping [8]:

$$R(T) = R_0 \times \exp\sqrt{\frac{T_0}{T}},$$

where  $R_0$  and  $T_0$  are constants largely determined by geometry and doping, respectively.



#### **Frame**

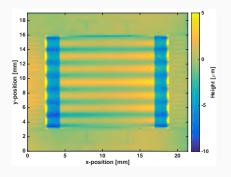
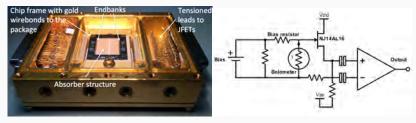


Figure 1: replace this with In bump SEM

- The chip frame is designed so that any two bolometer chips can be hybridized together.
- Large gold-covered areas serve as heat sinks.

Package and readout

# Package and readout - dark tests

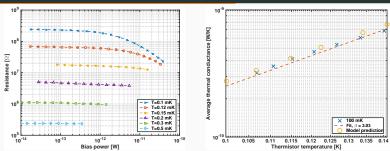


- Thermistor operates under current bias  $(R_{bias} >> R_{therm})$ .
- Bolometer is connected to a cryogenic (130 K) JFET amplifier with tensioned leads, mitigating capacitive microphonic contamination of the signal band. We use Interfet NJ14AL16 JFETs that are screened for low noise performance (5.5  $\rm nV/\sqrt{Hz}$  at 100 Hz).
- Amplifier converts the high source impedance of the thermistors (M $\Omega$ -scale) to the low output impedance of the JFETs (1.8 k $\Omega$ ).
- Low impedance signal is AC coupled to a room temperature amplifier.



**Detector performance** 

# Performance - load curves



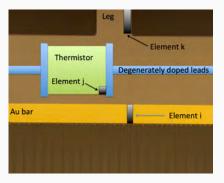
- Determine  $R_0$  and  $T_0$  from the measured resistances under low electrical bias.
  - ightarrow  $T_0=15.11$  K and  $R_0=911$   $\Omega.$  Operating resistance: 5.42 M $\Omega.$
- Determine average thermal conductance  $\bar{G}$  between the thermistors and the bath from the high-bias end of the load curves:

$$\bar{G} = \frac{P_{\text{bias}}}{T_1 - T_2}. ag{5}$$

- Fit to the measured  $\bar{G}$  with a function  $\tilde{G} = G_0 \times T^{\beta}$ .
  - $\rightarrow$  The fit is close to the expected value ( $\beta_{phonon} = 3$ ).



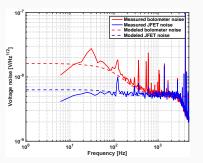
# Performance - thermal model

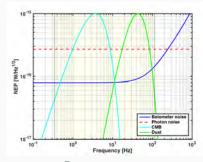


- For the endbank geometry, break Au bar, thermistors, and legs into small elements.
- Solve for the etendue  $A\Omega_{ij,ik}$  beween all elements.
- Heat flow between elements (e.g., between i and j) is given by  $P_{ij} = A_{ij} \left( T_i^{\nu} T_j^{\nu} \right)$ .
- Determine G between elements, determine C from material properties/geometries, measure VRH parameters R<sub>0</sub> and T<sub>0</sub>, and solve for non-equilibrium bolometer noise [9]:

$$NEP_{bolometer}^{2} = \gamma_{1}4k_{b}T^{2}G + \frac{1}{S^{2}}\left(\gamma_{2}4k_{b}TR + e_{n}^{2} + \gamma_{3}i_{n}^{2}R + \gamma_{4}NEP_{excess}^{2}\right).$$

# Performance - noise





- ullet Thermal model reproduces the measured  $ar{G}$  well.
- Modeled noise fits the measured noise well for multiple bias conditions.
- Running the model for the optical and electrical bias conditions expected during flight, we calculate a bolometer NEP of  $7.93\times10^{-17}~W/\sqrt{\rm Hz}.$



Expect to be photon noise limited across the entire PIXIE bandwidth.

# Conclusions

### **Conclusions**

- We designed, fabricated, and characterized large area polarization-sensitive bolometers for the PIXIE experiment.
- Mechanical characterization of the fabricated PIXIE bolometers shows that the tensioning scheme successfully flattens the absorber strings.
  - Enables indium bump hybridization of a pair of bolometer chips.
  - Mitigates microphonic sensitivity during launch.
- The dark data provide significant insight into the thermal behavior of the endbanks.
  - Thermal model agress well with the data.
  - The results indicate that the PIXIE bolometers satisfy the sensitivity and bandwidth requirements of the space mission.
- Upcoming work:
  - Characterize the absorber structure (dark measurements of thermal transport and AC impedance, optical measurements with a cryogenic blackbody source.)
  - Subject a hybridized pair of bolometers to environmental testing.

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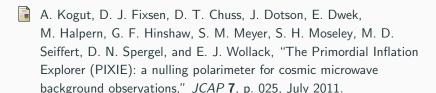


# Backup

Backup



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